

LECTURE 5

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Spot the difference.

Uncertainty – Error propagation Shallow water Up-down decomposition Anisotropy



Uncertainty – Error propagation



Propagation of uncertainty. Error propagation

Example with two variables Taylor to first order:

$$f(x,y) = f(x_0, y_0) + \frac{\partial f(x_0, y_0)}{\partial x}(x - x_0) + \frac{\partial f(x_0, y_0)}{\partial y}(y - y_0)$$

Or

$$f(x,y) - f(x_0, y_0) = \frac{\partial f(x_0, y_0)}{\partial x} (x - x_0) + \frac{\partial f(x_0, y_0)}{\partial y} (y - y_0)$$

$$\delta f(x,y) = \frac{\partial f(x_0, y_0)}{\partial x} \delta x + \frac{\partial f(x_0, y_0)}{\partial y} \delta y$$

Variation in function f as a function of variations in parameters x and y.



Have:

$$\delta f(x,y) = \frac{\partial f(x_0, y_0)}{\partial x} \delta x + \frac{\partial f(x_0, y_0)}{\partial y} \delta y$$

Suppose we do N measurements of f(x, y). For the n'th measurement:

$$\delta f_n = \frac{\partial f(x_0, y_0)}{\partial x} \delta x_n + \frac{\partial f(x_0, y_0)}{\partial y} \delta y_n$$

$$\delta f_n^2 = \left(\frac{\partial f}{\partial x}\right)^2 \delta x_n^2 + \left(\frac{\partial f}{\partial y}\right)^2 \delta y_n^2 + 2\frac{\partial f}{\partial x}\frac{\partial f}{\partial y}cov(x_n, y_n)$$



Have:

$$\delta f_n^2 = \left(\frac{\partial f}{\partial x}\right)^2 \delta x_n^2 + \left(\frac{\partial f}{\partial y}\right)^2 \delta y_n^2 + 2\frac{\partial f}{\partial x}\frac{\partial f}{\partial y}cov(x_n, y_n)$$

If independent variables:

$$\frac{\sum_{n=1}^{N} \delta f_n^2}{N} = \left(\frac{\partial f}{\partial x}\right)^2 \frac{\sum_{n=1}^{N} \delta x_n^2}{N} + \left(\frac{\partial f}{\partial y}\right)^2 \frac{\sum_{n=1}^{N} \delta y_n^2}{N}$$

In terms of standard deviations:

$$s_{f}^{2} = \left(\frac{\partial f}{\partial x}\right)^{2} s_{x}^{2} + \left(\frac{\partial f}{\partial y}\right)^{2} s_{y}^{2}$$

Will use notation of the form: $\delta f^{2} = \left(\frac{\partial f}{\partial x}\right)^{2} \delta x^{2} + \left(\frac{\partial f}{\partial y}\right)^{2} \delta y^{2}$



Have

$$\delta f^2 = \left(\frac{\partial f}{\partial x}\right)^2 \delta x^2 + \left(\frac{\partial f}{\partial y}\right)^2 \delta y^2$$

The expected uncertainty/error in maesuring f due to uncertainty/error in x and y:

$$\delta f = \sqrt{\left(\frac{\partial f}{\partial x}\right)^2 \delta x^2 + \left(\frac{\partial f}{\partial y}\right)^2 \delta y^2}$$

Generalized:

$$\delta f(\mathbf{x}) = \sqrt{\sum_{i} |\frac{\partial f(\mathbf{x})}{\partial x_{i}}|^{2} |\delta x_{i}|^{2}}$$



Speed trap: Measured fixed distance and interval time measurement



$$v = \frac{s_2 - s_1}{t_2 - t_1} = \frac{s}{t}$$



The distance from A to B is s = 100 m but there is an uncertainty related to measuring the distance $s : \delta s$

Likewise, there is an uncertainty related to measuring the time it takes to drive from A to B: δt

The velocity is:
$$v(s,t) = \frac{s}{t}$$

The uncertainty in the velocity measurement is:

Explicitly:
$$\delta v = \sqrt{\left(\frac{1}{t}\right)^2 \delta s^2 + \left(\frac{s}{t^2}\right)^2 \delta t^2}$$

$$\delta v = \sqrt{\left(\frac{\partial v}{\partial s}\right)^2} \,\delta s^2 + \left(\frac{\partial v}{\partial t}\right)^2 \,\delta t^2$$

Have

$$\delta v = \sqrt{\left(\frac{1}{t}\right)^2 \delta s^2 + \left(\frac{s}{t^2}\right)^2 \delta t^2}$$

Use
$$v = \frac{s}{t}$$
 to obtain:
$$\frac{\delta v}{v} = \sqrt{\left(\frac{\delta s}{s}\right)^2 + \left(\frac{\delta t}{t}\right)^2}$$

Suppose the task is to measure velocities up to 110 km/h with accuracy 1.1 km/h or better.

For the moment, assume perfect timing ($\delta t = 0$). Sufficient accuracy if:

$$\frac{\delta v}{v} \ge \frac{\delta s}{s}$$



Have for perfect time measurements:

$$\frac{\delta s}{s} \le 0.01$$

If the distance $s = 100 \text{ m} : \delta s \le 1 \text{ m}$

Assumption: $\delta s \approx 0.1$ m with laser

$$\frac{\delta v}{v} = \sqrt{\left(\frac{\delta s}{s}\right)^2 + \left(\frac{\delta t}{t}\right)^2} \longrightarrow \quad 0.01 \ge \sqrt{(0.001)^2 + \left(\frac{\delta t}{t}\right)^2}$$

Obtain

$$\frac{\delta t}{t} < 0.01$$

Good accuracy on distance measurement implies that allmost all potential uncertainty is related to time measurement



Have

$$\frac{\delta t}{t} < 0.01$$

Small time intervals from A to B will give largest uncertainty. This is for highest velocity.

Have 110 km/h \approx 30 m/s. Expected shortest time is t = 3.333 s

Acceptable uncertainty in time measurement, δt , is 0.033 s or 33 msec

Manual timing with stopwatch or electronic timing?







Assume observed inline electric field can be approximated by:

 $E_x(\boldsymbol{x}_r|\boldsymbol{x}_s) = G_{xn}^{EJ}(\boldsymbol{x}_r|\boldsymbol{x}_s)LJ_n\alpha + N$

 $G_{xn}^{EJ}(x_r|x_s)$: Electric field Green's function. Often named «The Earth's impulse response» Here it serves the role as an ideal response without errors or uncertainty.

- *L*: Length of electric dipole
- J_n : Strength of transmitted current
- α : Receiver calibration factor, nominal value is 1.0
- *N*: Additive noise. Can be receiver self noise, MT noise, swell noise, motion noise, ...

Uncertainty



Receiver:

- Position
- Direction
- Calibration
- Timing
- Self noise
- Motion noise turbulence
- Swell noise
- MT noise (Can be estimated/partly removed)

Transmitter:

- Front electrode position
- Aft electrode position
- Current measurement
- Timing

Effective length, feathering, pitch



Have

$$E_{x}(\boldsymbol{x}_{r}|\boldsymbol{x}_{s}) = G_{xn}^{EJ}(\boldsymbol{x}_{r}|\boldsymbol{x}_{s})LJ_{n}\alpha + N$$

For simplicity of derivation we assume a plane layer earth $\mathbf{x} = \mathbf{x}_r - \mathbf{x}_s$ $E_x(\mathbf{x}) = G_{xn}^{EJ}(\mathbf{x})LJ_n\alpha + N$

For the source components

$$J_{x} = J\cos(\varphi)\cos(\theta)$$

$$J_{y} = J\sin(\varphi)\cos(\theta)$$

$$J_{z} = J\sin(\theta)$$
Nominal $\varphi = 0$, $\theta = 0$
Perfect inline transmitter give: $J_{x} = J$, $J_{y} = 0$, $J_{z} = 0$

$$E_{x}(\boldsymbol{x}) \rightarrow E_{x}(\boldsymbol{p}) \qquad \boldsymbol{p} = [\boldsymbol{x}, \alpha, J, L, \theta, \varphi, N]^{T}$$

Have derived

$$\delta f(\mathbf{x}) = \sqrt{\sum_{i} |\frac{\partial f(\mathbf{x})}{\partial x_{i}}|^{2} |\delta x_{i}|^{2}}$$

For inline electromagnetic field:

$$\delta E_{x}(\boldsymbol{p}) = \sqrt{\sum_{i} |\frac{\partial E_{x}(\boldsymbol{p})}{\partial p_{i}}|^{2} |\delta p_{i}|^{2}}$$
$$E_{x}(\boldsymbol{p}) = G_{xn}^{EJ}(\boldsymbol{x})LJ_{n}\alpha + N$$
$$\boldsymbol{p} = [\boldsymbol{x}, \alpha, J, L, \theta, \varphi, N]^{T}$$

For simplicity of notation: $D = LJ\alpha$



Next step is to carry out calculation of partial derivatives with respect to parameter vector

$$E_{x}(\boldsymbol{p}) = G_{xn}^{EJ}(\boldsymbol{x})LJ_{n}\alpha + N \qquad \boldsymbol{p} = [\boldsymbol{x}, \alpha, J, L, \theta, \varphi, N]^{T} \qquad D = LJ\alpha$$

Spatial coordinates:

$$\partial_{x} E_{x}(\boldsymbol{p}) = \partial_{x} G_{xx}^{EJ}(\boldsymbol{x}) D$$
$$\partial_{y} E_{x}(\boldsymbol{p}) = 0$$
$$\partial_{z} E_{x}(\boldsymbol{p}) = \partial_{z} G_{xx}^{EJ}(\boldsymbol{x}) D$$

Directional coordinates:

$$J_{x} = J\cos(\varphi)\cos(\theta)$$

$$\partial_{\varphi}E_{x}(\boldsymbol{p}) = 0 \qquad \qquad J_{y} = J\sin(\varphi)\cos(\theta) \quad (G_{xy} = 0)$$

$$J_{z} = J\sin(\theta)$$

$$\partial_{\theta}E_{x}(\boldsymbol{p}) = G_{xz}^{EJ}(\boldsymbol{x})D$$



Next step is to carry out calculation of partial derivatives with respect to parameter vector

$$E_{x}(\boldsymbol{p}) = G_{xn}^{EJ}(\boldsymbol{x})LJ_{n}\alpha + N \qquad \boldsymbol{p} = [\boldsymbol{x}, \alpha, J, L, \theta, \varphi, N]^{T} \qquad D = LJ\alpha$$

Dipole moment-receiver calibration:

$$\partial_L E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{L}$$
$$\partial_J E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{J}$$
$$\partial_\alpha E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{\alpha}$$

Additive noise:

$$\partial_N E_x(\boldsymbol{p}) = 1$$



Partial uncertainties

$$\partial_{x} E_{x}(\boldsymbol{p}) = \partial_{x} G_{xx}^{EJ}(\boldsymbol{x}) D \longrightarrow \delta E_{x}(X) = |\partial_{x} G_{xx}^{EJ}(\boldsymbol{x}) D \delta x|$$
$$\partial_{z} E_{x}(\boldsymbol{p}) = \partial_{z} G_{xx}^{EJ}(\boldsymbol{x}) D \longrightarrow \delta E_{x}(Z) = |\partial_{z} G_{xx}^{EJ}(\boldsymbol{x}) D \delta z|$$

$$\partial_L E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{L}$$

$$\partial_J E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{J}$$

$$\delta E_x(C) = |G_{xx}^{EJ}(\boldsymbol{x}) D| \sqrt{\left(\frac{\delta L}{L}\right)^2 + \left(\frac{\delta J}{J}\right)^2 + \left(\frac{\delta \alpha}{\alpha}\right)^2}$$

$$\partial_\alpha E_x(\boldsymbol{p}) = G_{xx}^{EJ}(\boldsymbol{x}) D \frac{1}{\alpha}$$

 $\partial_{\theta} E_{x}(\boldsymbol{p}) = G_{xz}^{EJ}(\boldsymbol{x})D \longrightarrow \delta E_{x}(\theta) = |G_{xz}^{EJ}(\boldsymbol{x})D \,\delta\theta|$

 $\partial_N E_x(\mathbf{p}) = 1 \qquad \longrightarrow \qquad \delta E_x(\mathbf{N}) = |\Delta N|$



Have

$$\delta E_{x}(X) = |\partial_{x} G_{xx}^{EJ}(\mathbf{x}) D \, \delta x|$$

$$\delta E_{x}(Z) = |\partial_{z} G_{xx}^{EJ}(\mathbf{x}) D \, \delta z|$$

$$\delta E_{x}(C) = |G_{xx}^{EJ}(\mathbf{x}) D| \sqrt{\left(\frac{\delta L}{L}\right)^{2} + \left(\frac{\delta J}{J}\right)^{2} + \left(\frac{\delta \alpha}{\alpha}\right)^{2}}$$

$$\delta E_{x}(\theta) = |G_{xz}^{EJ}(\mathbf{x}) D \, \delta \theta|$$

Collect all terms that scales with Green's functions and dipole moment:

$$|\delta E_{x}(\mathbf{M})|^{2} = |\delta E_{x}(\mathbf{X})|^{2} + |\delta E_{x}(\mathbf{Z})|^{2} + |\delta E_{x}(\mathbf{C})|^{2} + |\delta E_{x}(\theta)|^{2}$$

The additive term:

 $\delta E_{x}(\mathbf{N}) = |\Delta N|$

The total uncertainty:

$$\delta E_{x}(\boldsymbol{p}) = \sqrt{\sum_{i} |\frac{\partial E_{x}(\boldsymbol{p})}{\partial p_{i}}|^{2} |\delta p_{i}|^{2}} \longrightarrow \delta E_{x}(\boldsymbol{p}) = \sqrt{|\delta E_{x}(\mathbf{M})|^{2} + |\delta E_{x}(\mathbf{N})|^{2}}$$



Plot color coding

$$\delta E_{x}(X) = |\partial_{x} G_{xx}^{EJ}(x) D \, \delta x|$$

$$\delta E_{x}(Z) = |\partial_{z} G_{xx}^{EJ}(x) D \, \delta z|$$

$$\delta E_{x}(C) = |G_{xx}^{EJ}(x) D| \sqrt{\left(\frac{\delta L}{L}\right)^{2} + \left(\frac{\delta J}{J}\right)^{2} + \left(\frac{\delta \alpha}{\alpha}\right)^{2}}$$

$$\delta E_{x}(\theta) = |G_{xz}^{EJ}(x) D \, \delta \theta|$$

$$\delta E_{x}(N) = |\Delta N|$$

$$\delta E_{x}(n) = \sqrt{|\Delta E_{x}(M)|^{2} + |\Delta E_{x}(N)|^{2}}$$

 $\delta E_{\chi}(\boldsymbol{p}) = \sqrt{|\delta E_{\chi}(\mathbf{M})|^2 + |\delta E_{\chi}(\mathbf{N})|^2}$



Scattered fields

Misfit in first iteration:

$$\Delta E_x^0(\boldsymbol{p}) = |E_x^{Obs}(\boldsymbol{p}) - E_x^0(\boldsymbol{x}, \omega)|$$

Assume that 67 percent (2/3) of transverse resistance recovered at iteration n:

$$\Delta E_{x}^{n}(\boldsymbol{p}) = |E_{x}^{Obs}(\boldsymbol{p}) - E_{x}^{n}(\boldsymbol{x},\omega)|$$

True model

Start model

Partially recovered model after n iterations





Inversion

L1 inversion data misfit kernel:

$$\Psi^n(\boldsymbol{p}) = \frac{\Delta E_x^n(\boldsymbol{p})}{\delta E_x(\boldsymbol{p})}$$

Inspect ratio of residual misfit field to uncertainty:

Hard to extract more resistivity information if residual data misfit is of same magnitude as uncertainty. Critical value for Ψ is 1. Further iterations make sense if Ψ larger than unity

Usually L2 inversion data misfit kernels used:

$$\varepsilon = \sum_{Observations} \left(\frac{\Delta E_x^n}{\delta E_x} \right)^2$$





Resistivity model

f=0.25 Hz





Depth:2000 m Cur:1000 A Dx:15.0 m Dz:5.0 m Dcal:0.01 Dcur:0.02 Gr:A BI:Dx Cy:Dz Gy:Tlt R%:67.0

















Depth:2000 m Cur:1000 A Dx:15.0 m Dz:5.0 m Dcal:0.01 Dcur:0.02 Gr:A BI:Dx Cy:Dz Gy:Tlt R%:67.0













For

$$\delta E_{x}(C) = |G_{xx}^{EJ}(\boldsymbol{x})D| \sqrt{\left(\frac{\delta L}{L}\right)^{2} + \left(\frac{\delta J}{J}\right)^{2} + \left(\frac{\delta \alpha}{\alpha}\right)^{2}}$$

Assume in the following:

$$\frac{\delta L}{L} = \frac{\delta J}{J} = \frac{\delta \alpha}{\alpha}$$
$$\delta E_x(C) = |G_{xx}^{EJ}(\mathbf{x})D| \frac{\delta A}{A}$$

For example

$$\frac{\delta A}{A} = \sqrt{3} \frac{\delta L}{L} \approx 1.7 \frac{\delta L}{L}$$







$$\Delta E_x^0(\boldsymbol{p}) = |E_x^{Obs}(\boldsymbol{p}) - E_x^0(\boldsymbol{x}, \omega)|$$
$$\Delta E_x^n(\boldsymbol{p}) = |E_x^{Obs}(\boldsymbol{p}) - E_x^n(\boldsymbol{x}, \omega)|$$
$$\delta E_x(\boldsymbol{p}) = \sqrt{|\delta E_x(\mathbf{M})|^2 + |\delta E_x(\mathbf{N})|^2}$$



Typical accuracy as of 2010

 $\delta x = 15 \text{ m}$ $\delta z = 5 \text{ m}$ $\delta \theta = 1^{\circ}$



Spot the difference.



Target down

Typical accuracy as of 2010

 $\delta x = 15 \text{ m}$ $\delta z = 5 \text{ m}$ $\delta \theta = 1^{\circ}$





Target down

 $\delta x = 15 \text{ m}$ $\delta z = 5 \text{ m}$ $\delta \theta = 1^{\circ}$

Receiver noise 10^{-11} V/m






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Target down

$$\delta x = 5 \text{ m}$$

 $\delta z = 2 \text{ m}$
 $\delta \theta = 0.4^{\circ}$

Receiver noise 10^{-11} V/m Transmitter current 10 kA Better calibration and δL Better navigation x & zBetter navigation θ





 $\delta x = 5 \text{ m}$ $\delta z = 2 \text{ m}$ $\delta \theta = 0.4^{\circ}$

Receiver noise 10^{-11} V/m Transmitter current 10 kA Better calibration and δL Better navigation x & zBetter navigation θ



Shallow water – 40 m

Problem is MT – swell – motion noise

$\delta x = 1$	1 m
$\delta z = 1$	1 m
$\delta\theta =$	0°

Receiver noise 10^{-9} V/m



Have

$$\delta E_x(\boldsymbol{p}) = \sqrt{|\delta E_x(\mathbf{M})|^2 + |\delta E_x(\mathbf{N})|^2}$$
$$\delta E_x(\mathbf{M})|^2 = |\delta E_x(\mathbf{X})|^2 + |\delta E_x(\mathbf{Z})|^2 + |\delta E_x(\mathbf{C})|^2 + |\delta E_x(\boldsymbol{\theta})|$$
$$\Delta E_x^0(\boldsymbol{p}) = |E_x^0|^{\delta E_x}(\boldsymbol{p}) - E_x^0|^{\delta E_x}(\boldsymbol{\theta})|$$

$$\Delta E_x^0(\boldsymbol{p}) = |E_x^{ODS}(\boldsymbol{p}) - E_x^0(\boldsymbol{x}, \omega)|$$

Let *x* here mean source-receiver offset

$$\gamma(x) = \frac{\delta E_x(x|\mathsf{M})}{E_x^{Obs}(x)}$$

Can write:

$$\delta E_x(x) = \sqrt{|\gamma(x)E_x^{Obs}(x)|^2 + \Delta N^2}$$

How does γ behave as a function of source-receiver offset?



2



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Quick and dirty estimate of uncertainty:

$$\delta E_x(x) \approx \sqrt{\gamma^2 |E_x^{Obs}(x)|^2 + \Delta N^2}$$

Even dirtier estimate of uncertainty: $\delta E_x(x) \approx \gamma |E_x^{Obs}(x)| + \Delta N$



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Frequency vs. Offset

In order to find the best fitting range of frequencies for a survey it is important to find a waveform with frequencies at and around the peak sensitivity for a given target.

At the same time one should keep in mind that different frequencies have different penetration and resolution.



Shallow water















 Air		
Water	40 - 2000 m	0.3 Ohm-m
Formation	1000 m	2 Ohm-m
 Resistor	50 m	50 Ohm-m
Formation		2 Ohm-m

All examples are for 0.25 Hz Results not particular for that frequency



Air		А
Water	40 - 2000 m	
Formation	1000 m	
Resistor	50 m	
Formation		

Air		В
Water	40 - 2000 m	
Formation		

Full waveform modeling of the scattered field from a thin resistor

$$\Delta E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s) = E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s, A) - E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s, B)$$

$$E_{\chi\chi}(\boldsymbol{x}_r|\boldsymbol{x}_s,\mathsf{A}) = E_{\chi\chi}(\boldsymbol{x}_r|\boldsymbol{x}_s,\mathsf{B}) + \Delta E_{\chi\chi}(\boldsymbol{x}_r|\boldsymbol{x}_s)$$









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Spot the difference.



Shallow water CSEM very difficult if scattered field same amplitude for all waterdepths







The amplitude of the scattered field increase significantly in waterdepths less than 300 m









Scattered fields normalized on the 2000 m waterdepth case

Resistor burial depth 1000 m

Resistor burial depth 3000 m

Enhanced scattered-field effect is not restricted to a particular burial depth or frequency

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Spot the difference.



Scattered field of same magnitude as background field for a fairly large offset interval.

Marine CSEM in shallow water feasible.



Magnitude of airwave increase as waterdepth is reduced

The response from a thin resistive layer increase as waterdepth is reduced

The increase in the response from a thin resistive layer is sufficiently strong to make marine CSEM in shallow water feasible





Mittet and Morten 2012: Error propagation analysis to estimate uncertainty in observation

 $E_{xx}^{Obs}(\boldsymbol{x}_r|\boldsymbol{x}_s; L, J, \beta, N,) \approx G_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s) \sqcup \beta + N$

$$\delta E_{xx}^{Obs}(\boldsymbol{p}) = \sqrt{\sum_{i} |\frac{\partial E_{xx}^{Obs}(\boldsymbol{p})}{\partial p_{i}}|^{2} |\Delta p_{i}|^{2}}$$

Contributions to uncertainty are both multiplicative and additive

For model used here we find that multplicative contributions approximately constant with offset (Offset > 2 km)

Simplified model for the uncertainty in the observed data:

$$\delta E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s) = \sqrt{\alpha^2 |E_{xx}^{Obs}(\boldsymbol{x}_r|\boldsymbol{x}_s)|^2 + \eta^2}$$



Scattered (≈misfit) field from full waveform modeling:

 $\Delta E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s) = E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s, \mathsf{A}) - E_{xx}(\boldsymbol{x}_r|\boldsymbol{x}_s, \mathsf{B})$

Uncertainty in the observed field:

$$\delta E_{xx}(\boldsymbol{x}_r | \boldsymbol{x}_s) = \sqrt{\alpha^2 |E_{xx}^{Obs}(\boldsymbol{x}_r | \boldsymbol{x}_s)|^2 + \eta^2}$$

L1 inversion kernel at first iteration:

$$\Psi(\boldsymbol{x}_r | \boldsymbol{x}_s) = \frac{|\Delta E_{xx}(\boldsymbol{x}_r | \boldsymbol{x}_s)|}{|\delta E_{xx}(\boldsymbol{x}_r | \boldsymbol{x}_s)|}$$

L2 inversion kernel at first iteration:

$$\Psi(\boldsymbol{x}_r | \boldsymbol{x}_s) = \frac{|\Delta E_{xx}(\boldsymbol{x}_r | \boldsymbol{x}_s)|^2}{|\delta E_{xx}(\boldsymbol{x}_r | \boldsymbol{x}_s)|^2}$$
(Tarantola, 1984)



Noise models

$$\delta E_{xx}(x_r | x_s) = \sqrt{\alpha^2 |E_{xx}^{Obs}(x_r | x_s)|^2|} + \eta^2$$

Based on error propagation analysis
 $\alpha = 0.03$
Based on real data
 $\eta(2000 \text{ m}) = 5 \times 10^{-16} \frac{\text{V}}{\text{Am}^2}$
 $\eta(300 \text{ m}) = 3 \times 10^{-15} \frac{\text{V}}{\text{Am}^2}$
 $\eta(100 \text{ m}) = 6 \times 10^{-15} \frac{\text{V}}{\text{Am}^2}$
 $\eta(40 \text{ m}) = 1.5 \times 10^{-14} \frac{\text{V}}{\text{Am}^2}$





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Spot the difference.



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Spot the difference.



From 2000 m – 400 m: Increase in airwave and additive noise give reduced sensitivity

From 400 m – 40 m : Increase in scattered field balance the airwave effect



Up-down decomposition





Before up-down decomposition





After up-down decomposition







The purpose of U-D decomposition is to reduce the contribution from «large amplitude» downgoing field components like the airwave and MT fields

After decomposition further processing is performed on the upgoing field that has interacted with the subsurface

Maxwell equations for 1D MT

$$\begin{bmatrix} J_x^s + \sigma E_x \\ J_y^s + \sigma E_y \\ 0 + \sigma E_z \end{bmatrix} = \begin{bmatrix} -\partial_z H_y \\ \partial_z H_x \\ 0 \end{bmatrix} \qquad \begin{bmatrix} i\omega\mu_0 H_x \\ i\omega\mu_0 H_y \\ i\omega\mu_0 H_z \end{bmatrix} = \begin{bmatrix} -\partial_z E_y \\ \partial_z E_x \\ 0 \end{bmatrix}$$

Obtain two sets of equations that describe two different polarizations:

$$\partial_z H_y + \sigma E_x = -J_x^s \qquad \qquad \partial_z H_x + \sigma E_y = -J_y^s \\ \partial_z E_x - i\omega\mu_0 H_y = 0 \qquad \qquad \partial_z E_y + i\omega\mu_0 H_x = 0$$

Equations for both polarizations :

 $\partial_z^2 E_x + i\omega\mu_0 \sigma E_x = -i\omega\mu_0 J_x^s$ $\partial_z^2 E_y + i\omega\mu_0 \sigma E_y = -i\omega\mu_0 J_y^s$

Sufficient to concentrate on x-polarization to understand the physics.







$$\begin{aligned} \partial_{z} E_{x} - i\omega\mu_{0}H_{y} &= 0 \\ \\ E_{x}^{D}(z,\omega) &= E_{x}(z_{a},\omega)e^{ik_{\omega}(z-z_{a})} \end{aligned} \qquad \begin{aligned} E_{x}^{U}(z,\omega) &= E_{x}(z_{b},\omega)e^{ik_{\omega}(z_{b}-z)} \\ \\ E_{x}^{U}(z,\omega) &= E_{x}(z_{b},\omega)e^{ik_{\omega}(z_{b}-z)} \end{aligned} \qquad \begin{aligned} k_{\omega} &= \sqrt{i\omega\mu_{0}\sigma} \end{aligned}$$



Have in general: $\partial_z E_x = i\omega\mu_0 H_y$



$$\begin{array}{l} \partial_{z}E_{x}-i\omega\mu_{0}H_{y}=0\\ \\ E_{x}^{D}(z,\omega)=E_{x}(z_{a},\omega)e^{ik_{\omega}(z-z_{a})}\\ \end{array} \quad \begin{array}{l} E_{x}^{U}(z,\omega)=E_{x}(z_{b},\omega)e^{ik_{\omega}(z_{b}-z)}\\ \\ E_{x}^{U}(z,\omega)=E_{x}(z_{b},\omega)e^{ik_{\omega}(z-z_{a})}\\ \end{array} \quad \begin{array}{l} k_{\omega}=\sqrt{i\omega\mu_{0}\sigma}\\ \end{array}$$



Have in general: $\partial_z E_x = i\omega\mu_0 H_v$ Assume downgoing field only at *z*: $E_x^D(z,\omega) = \frac{\omega\mu_0}{k_\omega} H_y^D(z,\omega)$ $E_x^D(z,\omega) = \frac{\omega\mu_0}{\sqrt{i\omega\mu_0\sigma}} H_y^D(z,\omega)$ $E_{\chi}^{D}(z,\omega) = ZH_{\chi}^{D}(z,\omega)$

$$\begin{array}{l} \partial_{z}E_{x}-i\omega\mu_{0}H_{y}=0\\ \\ E_{x}^{D}(z,\omega)=E_{x}(z_{a},\omega)e^{ik_{\omega}(z-z_{a})}\\ \end{array} \quad \begin{array}{l} E_{x}^{U}(z,\omega)=E_{x}(z_{b},\omega)e^{ik_{\omega}(z_{b}-z)}\\ \\ E_{x}^{U}(z,\omega)=E_{x}(z_{b},\omega)e^{ik_{\omega}(z-z_{a})}\\ \end{array} \quad \begin{array}{l} k_{\omega}=\sqrt{i\omega\mu_{0}\sigma}\\ \end{array}$$



Have in general: $\partial_z E_x = i\omega\mu_0 H_v$ Assume upgoing field only at *z*: $E_{x}^{U}(z,\omega) = -\frac{\omega\mu_{0}}{k_{\omega}}H_{y}^{U}(z,\omega)$ $E_{x}^{U}(z,\omega) = -\frac{\omega\mu_{0}}{\sqrt{i\omega\mu_{0}\sigma}}H_{y}^{U}(z,\omega)$ $E_{x}^{U}(z,\omega) = -ZH_{v}^{U}(z,\omega)$

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Characteristic impedance:

$$\mathbf{Z} = \frac{\omega\mu_0}{\sqrt{i\omega\mu_0\sigma}}$$

 $\mathbf{Z} = \sqrt{-i\omega\mu_0\rho}$

Characteristic impedance is a medium property and is completely determined by the medium resistivity (or conductivity).

Must not be confused by the expression for field impedance Z_{xy} used in MT processing:

$$Z_{xy} = \frac{E_x}{H_y}$$


We measure the total fields:

$$E_{x}(z,\omega) = E_{x}^{D}(z,\omega) + E_{x}^{U}(z,\omega)$$

$$H_{y}(z,\omega) = H_{y}^{D}(z,\omega) + H_{y}^{U}(z,\omega)$$

$$E_{x}^{D}(z,\omega) = ZH_{y}^{D}(z,\omega)$$
$$E_{x}^{U}(z,\omega) = -ZH_{y}^{U}(z,\omega)$$

$$E_{\chi}^{D}(z,\omega) = ZH_{y}^{D}(z,\omega) \longrightarrow E_{\chi}(z,\omega) - E_{\chi}^{U}(z,\omega) = ZH_{y}(z,\omega) - ZH_{y}^{U}(z,\omega)$$
$$E_{\chi}(z,\omega) - E_{\chi}^{U}(z,\omega) = ZH_{y}(z,\omega) + E_{\chi}^{U}(z,\omega)$$

Upgoing and downgoing fields are calculated from the measured fields:

$$E_{\chi}^{U}(z,\omega) = \frac{1}{2} [E_{\chi}(z,\omega) - ZH_{y}(z,\omega)]$$

$$E_{\chi}^{D}(z,\omega) = \frac{1}{2} [E_{\chi}(z,\omega) + ZH_{\gamma}(z,\omega)]$$

For vertically traveling field:

$$E_x^U(z,\omega) = \frac{1}{2} [E_x(z,\omega) - ZH_y(z,\omega)]$$

General solution:

$$E_{\chi}^{U}(z,\omega) = \frac{1}{2} \left[E_{\chi}(z,\omega) - Z \frac{(k_{\chi}k_{y}H_{\chi}(z,\omega) + (k_{\omega}^{2} - k_{\chi}^{2})H_{y}(z,\omega))}{k_{\omega}\sqrt{k_{\omega}^{2} - k_{\chi}^{2} - k_{y}^{2}}} \right]$$

Vertical propagation $k_x = k_y = 0$

Practical Up-Down decomposition is performed with: $E_x^U(z,\omega) = \frac{1}{2} [E_x(z,\omega) - ZH_y(z,\omega)]$



Electric fields paralell to an interface are contineous over interface.

Current normal to interfaces is contineous

Magnetic fields continous if non-magnetic material



Electric fields paralell to an interface are contineous over interface. Magnetic fields paralell to an interface are contieous over interface.





















Separation above and below seabed























Spot the difference.















$$E_{\chi}^{U}(z,\omega) = \frac{1}{2} \left[E_{\chi}(z,\omega) - Z \frac{(k_{\chi}k_{y}H_{\chi}(z,\omega) + (k_{\omega}^{2} - k_{\chi}^{2})H_{y}(z,\omega))}{k_{\omega}\sqrt{k_{\omega}^{2} - k_{\chi}^{2} - k_{y}^{2}}} \right]$$







$$E_{\chi}^{U}(z,\omega) = \frac{1}{2} [E_{\chi}(z,\omega) - Z_{W}H_{y}(z,\omega)]$$









































Doing Up-Down decomposition is not the same as doing a deep water experiment

Air Water Overburden Reservoir

Internal multiples in waterlayer is not removed.





Spot the difference.

🕿 emgs



Spot the difference.

emgs



emgs

Spot the difference.



emgs

Spot the difference.





20 (m) 1210 (D) Resistivity (D)











Anisotropy





Electrical anisotropy

- Resistivity within a formation is different in the vertical and horizontal directions.
- Reasons for this:
- Lithology, layering, grain orientation
- Fractures
- Diagenesis



Anisotropy Factor= ρ_v / ρ_h Values range from basin to basin and stratigraphic intervals.

Electrical anisotropy

A formation is said to be electrically anisotropic if its conductivity is direction dependendent.



Principle causes of anisotropy are: Lamination and bedding, grain shape and alignment, and fracturing.

TIV is typical for a formation with horizontal bedding and grain alignment.

General anisotropy is typical for a dipping formation.



In CSEM, it is most common to work with a TIV model.

<u>Good to know</u>

• TIV stands for "transverse isotropy with respect to a vertical axis of rotational symmetry".

In vertical wells, resistivity log



Electrical ansiotropy and resolution

- CSEM is a low-frequency technique, so we cannot hope to resolve conductivity variations on a scale similar to well log resistivity measurements.
- All we can expect is to measure a bulk conductivity of a rock slab with dimensions on the order of several meters.
- The bulk conductivity is, however, determined by the fine-scale structure and constituents of the slab.
- Material averaging laws dictate that the bulk conductivity is anisotropic even if the constituents are isotropic.







Spot the difference.

Material averaging for a formation with horizontal bedding



emgs

Spot the difference.



SPOT THE DIFFERENCE

Thank you